

Power Control and Achievable Rates for Fading CDMA: Generalized Iterative Waterfilling

Onur Kaya Sennur Ulukus
Department of Electrical and Computer Engineering
University of Maryland, College Park, MD 20742
onurkaya@eng.umd.edu ulukus@eng.umd.edu

Abstract

We characterize the optimum power control policies that achieve arbitrary rate tuples on the boundary of the capacity region of a power controlled, code division multiple access (CDMA) system in a fading channel. We propose a “generalized” waterfilling approach, and provide an iterative algorithm that solves for the optimum power allocation policy, for a given arbitrary rate tuple on the boundary of the capacity region.

I. INTRODUCTION

The capacity limits of communication systems subject to fading has recently drawn significant attention, and in the last decade, several results regarding the information theoretic capacities of many channel models have been reported. The particularly interesting types of channel models are those where the transmitter(s) and receiver(s) are able to track the variations in the channel, and therefore are capable of allocating the system resources and adapting their coding and decoding strategies to the variations in the channel, in order to improve the capacity. In this paper, we consider the uplink of a fading CDMA channel, and the resources we allocate are the available transmit powers.

The problem of power allocation in order to maximize the information theoretic capacity in the presence of fading was first studied for a single user channel in [1], where it was shown that, subject to an average power constraint and under the ergodicity assumption on the fading process, the ergodic capacity of the channel is maximized by allocating the total power of the user according to a waterfilling strategy, where the user “waterfills” its power in time, over the inverse of the channel states.

For multiple access channels, the capacity region is defined as the set of achievable rate tuples. For a scalar multiple access channel (MAC), [2] solved the power allocation problem with the goal of achieving a special rate tuple on the capacity region, the one that achieves the ergodic *sum* capacity. There, it was shown that in order to achieve the sum capacity, only the strongest user may transmit at any given time, and the optimum power control policy is again waterfilling, over disjoint sets of channel states.

The entire capacity region, and the corresponding power control policies for the scalar MAC were characterized in [3]. The capacity region is shown to be a union of the capacity regions (polymatroids) achievable by all valid power allocation policies,

i.e., the policies that satisfy the average power constraints. The optimal power allocation policy for each rate tuple on the capacity region is obtained by a greedy algorithm, which compares certain marginal utility functions, and makes use of the generalized symmetry properties of the rank function of the polymatroid corresponding to the rate tuple in question.

The capacity region for a non-fading vector MAC, where the total average power of the components of the transmitted vectors are constrained, is given by [4]. There, also an iterative waterfilling algorithm which allocates the powers over the components of the transmitted vector in order to maximize the *sum* capacity was proposed. The power allocation problem for a fading vector MAC was considered in [5], again with the aim of maximizing the sum capacity. It was shown that, the optimal power allocation in the fading case as well satisfies the Karush-Kuhn-Tucker (KKT) conditions, which can also be interpreted as simultaneous waterfilling, where the water levels are matrices.

In [6], the capacity region of a power controlled fading CDMA channel with perfect channel state information at the transmitters and the receiver was obtained. Also, the power allocation policy that achieves the sum capacity point on the capacity region boundary was found, and it was shown that the optimal power allocation policy is a simultaneous waterfilling of powers over the inverse of the SIRs the users would obtain if they transmitted with unit powers. It was also shown that, a one-user-at-a-time iterative waterfilling algorithm can be used to solve these simultaneous waterfilling conditions, and therefore to obtain the optimal power distributions of all users over all fading states.

In this paper, we consider the problem of solving for the power allocation policy that achieves an arbitrary rate tuple on the capacity region of fading CDMA. As in [3] and [4], this problem is equivalent to a maximization of a weighted sum of rates, subject to average power constraints. We make use of the concavity of the objective function and the convexity of the constraints, and write the KKT conditions at each fading state, for a given set of weights. We then develop a “generalized” waterfilling approach, where we gradually pour some power at some or all channel states until all the KKT conditions are satisfied. Using this approach, we propose a one-user-at-a-time algorithm similar to that in [6], [7], and show that it converges to the optimum power allocation for any given point on the boundary of the capacity region. This algorithm, while providing a systematic solution to the capacity achieving power allocation problem in fading CDMA, also provides as a special case, an

intuitive approach to the power allocation for scalar MAC in [3].

II. PROBLEM DEFINITION

We consider a symbol synchronous CDMA system with processing gain N , where all K users transmit to a single receiver site. In the presence of fading and AWGN, the received signal is given by [8],

$$\mathbf{r} = \sum_{i=1}^K \sqrt{p_i h_i} x_i \mathbf{s}_i + \mathbf{n} \quad (1)$$

where, for user i , x_i denotes the information symbol with $E[x_i^2] = 1$, $\mathbf{s}_i = [s_{i1}, \dots, s_{iN}]^\top$ denotes the unit energy signature sequence, $\sqrt{h_i}$ denotes the random and continuously distributed channel gain, and p_i denotes the transmit power; \mathbf{n} is a zero-mean Gaussian random vector with covariance $\sigma^2 \mathbf{I}_N$. We assume that the receiver and all of the transmitters have perfect knowledge of the channel states of all users represented as a vector $\mathbf{h} = [h_1, \dots, h_K]^\top$, and the components of \mathbf{h} are independent. We further assume that although the fading is slow enough to ensure constant channel gain in a symbol interval, it is fast enough so that within the transmission time of a block of symbols the long term ergodic properties of the fading process can be observed [9].

For the CDMA system given by (1), let the transmitters be able to choose their powers as a function of the channel states, subject to the average power constraints $E_{\mathbf{h}}[p_i(\mathbf{h})] \leq \bar{p}_i$. The following theorem from [6] gives the set of long term achievable rates, i.e., the capacity region, for fading CDMA.

Theorem 1: ([6]) The capacity region of a fading CDMA channel under additive white Gaussian noise, where users have perfect channel state information (CSI) and allocate their powers as a function of the CSI subject to average power constraints $E_{\mathbf{h}}[p_i(\mathbf{h})] \leq \bar{p}_i$ is given by,

$$\bigcup_{\{\mathbf{p}(\mathbf{h}): E_{\mathbf{h}}[p_i(\mathbf{h})] \leq \bar{p}_i, \forall i\}} \left\{ \mathbf{R} : \sum_{i \in \Gamma} R_i \leq E_{\mathbf{h}} \left[\frac{1}{2} \log \left| \mathbf{I}_N + \sigma^{-2} \sum_{i \in \Gamma} h_i p_i(\mathbf{h}) \mathbf{s}_i \mathbf{s}_i^\top \right| \right], \forall \Gamma \subset \{1, \dots, K\} \right\} \quad (2)$$

Figure 1 illustrates a typical capacity region for some fixed signature sequences \mathbf{s}_1 and \mathbf{s}_2 in a two user setting. Each of the pentagons corresponds to a valid power allocation policy. We have shown in [6] that the capacity region for fading CDMA is in general not strictly convex, and there may be a flat portion on the boundary of the capacity region, which coincides with the dominant face of the capacity region corresponding to the sum capacity maximizing power control policy. Now, note that, for any given point (R_1^*, R_2^*) on the boundary of the capacity region, one can find non-negative numbers μ_1 and μ_2 , such that the line $\mu_1 R_1 + \mu_2 R_2 = C$ is tangent to the capacity region for some $C = C^*(\mu_1, \mu_2)$, and in fact $C^*(\mu_1, \mu_2)$ is the maximum achievable value of $\mu_1 R_1 + \mu_2 R_2$. Therefore, the problem of finding the power control policy that corresponds to the rate pair

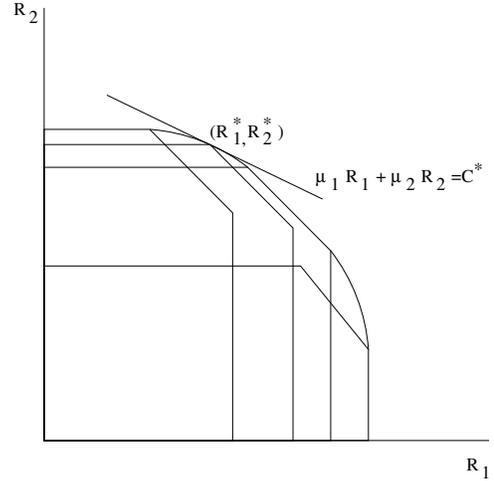


Fig. 1. Sample two user capacity region.

(R_1^*, R_2^*) is equivalent to maximizing $\mu_1 R_1 + \mu_2 R_2$ subject to the average power constraints. Here, μ_i s can be interpreted as the priorities assigned to each user. The desired rate pair (R_1^*, R_2^*) is either the corner of one of the pentagons specified by a power allocation policy as in (2), or it lies on one of the flat portions. If it is a corner, its coordinates can be written as a function of the power allocation policy using (2), and the maximization can be carried out. The case where (R_1^*, R_2^*) lies on one of the flat portions correspond to either the rather easier case where we want to maximize the sum capacity, which is solved in [6], [7], or the trivial case where one of the μ_i s is zero, and the problem reduces to a single user problem.

Having introduced the reasoning in the simple two user case, we now define our problem in the general K user case. Without loss of generality, assume $\mu_K > \dots > \mu_1$. Then, the optimum power allocation policy for $\{\mu_i\}_{i=1}^K$ is the solution to the maximization problem,

$$\begin{aligned} \max_{\mathbf{p}(\mathbf{h})} \quad & \frac{1}{2} E_{\mathbf{h}} \left[\mu_1 \log |\mathbf{I}_N + \sigma^{-2} \mathbf{S} \mathbf{D}(\mathbf{h}) \mathbf{S}^\top| \right. \\ & \left. + \sum_{i=2}^K (\mu_i - \mu_{i-1}) \log |\mathbf{I}_N + \sigma^{-2} \mathbf{S}_{E_i} \mathbf{D}_{E_i}(\mathbf{h}) \mathbf{S}_{E_i}^\top| \right] \\ \text{s.t.} \quad & E_{\mathbf{h}}[p_i(\mathbf{h})] \leq \bar{p}_i, \quad i = 1, \dots, K \\ & p_i(\mathbf{h}) \geq 0, \quad \forall \mathbf{h}, \quad i = 1, \dots, K \end{aligned} \quad (3)$$

where $\mathbf{S} = [\mathbf{s}_1 \dots \mathbf{s}_K]$, $\mathbf{D}(\mathbf{h}) = \text{diag}[p_1(\mathbf{h})h_1, \dots, p_K(\mathbf{h})h_K]$, $E_i = \{i, \dots, K\}$ and $\mathbf{p}(\mathbf{h}) = [p_1(\mathbf{h}), \dots, p_K(\mathbf{h})]$. Here, \mathbf{D}_{E_i} and \mathbf{S}_{E_i} refer to sub-matrices containing only the received powers and signature sequences of the users in the subset E_i . Note that, this is the fading CDMA version of equation (3) in [4], and is similar to equation (17) for the scalar case in [3].

III. GENERALIZED ITERATIVE WATERFILLING

Let us denote the objective function in (3) by $C_{\underline{\mu}}(p_1(\mathbf{h}), \dots, p_K(\mathbf{h}))$, where $\underline{\mu} = [\mu_1, \dots, \mu_K]$. In order to solve (3), we first note that the objective function is concave

in the power vector $\mathbf{p}(\mathbf{h})$, and further, it is strictly concave in the individual components $p_i(\mathbf{h})$ of $\mathbf{p}(\mathbf{h})$. The constraint set is convex (in fact, affine). Therefore, the solution to the maximization problem in (3) should satisfy the extended KKT conditions, which can be shown to reduce to,

$$\sum_{i=1}^k \frac{\mu_i - \mu_{i-1}}{a_{ki}(\mathbf{h}) + p_k(\mathbf{h})} \leq \lambda_k, \quad \forall \mathbf{h}, \quad k = 1, \dots, K \quad (4)$$

where, we have defined $\mu_0 \triangleq 0$ for notational convenience. Here, $a_{ki}(\mathbf{h})$ for $i \leq k \leq K$ is given by,

$$a_{ki}(\mathbf{h}) = \frac{1}{\sigma^{-2} \mathbf{h}_k \mathbf{s}_k^\top \left(\mathbf{I}_N + \sigma^{-2} \sum_{j=i, j \neq k}^K p_j(\mathbf{h}) \mathbf{h}_j \mathbf{s}_j \mathbf{s}_j^\top \right)^{-1} \mathbf{s}_k} \quad (5)$$

and denotes the inverse of the SIR user k would obtain at the output of an MMSE filter if it transmitted with unit power, when users $i, i+1, \dots, K$ are active. The condition in (4) is satisfied with equality at some \mathbf{h} , if $p_k(\mathbf{h}) > 0$. Since the optimum power allocation policy for a given $\underline{\mu}$ should simultaneously satisfy all the conditions given by (4), and the optimum power of each user k at each fading state \mathbf{h} depends on the power allocations of all other users at that state through $a_{ki}(\mathbf{h})$, it is hard to analytically solve for the optimum policy from the KKT conditions. Therefore, to proceed, we devise an iterative algorithm. Consider optimizing the power of *only* user k over all channel states, given the powers of all other users at all channel states,

$$\begin{aligned} p_k^{n+1} &= \arg \max_{p_k} C_{\underline{\mu}}^k(p_1^{n+1}, \dots, p_{k-1}^{n+1}, p_k, p_{k+1}^n, \dots, p_K^n) \\ &= \arg \max_{p_k} C_{\underline{\mu}}^k(p_k) \end{aligned} \quad (6)$$

where $C_{\underline{\mu}}^k(p_k)$ denotes the first k terms that contain contributions from user k to $C_{\underline{\mu}}(\mathbf{p}(\mathbf{h}))$.

The convergence of such an algorithm has been proved for the case of sum capacity in [6], [7] for fading channels, and in [4] for non-fading channels. The objective function here satisfies the same concavity and strict concavity properties as the sum capacity, and the constraint set is the same as in [6], [7]; therefore the proof in [6], [7] immediately applies to the case of unequal μ_i s here, and the update (6) converges to the optimal power allocation. Thus, it is sufficient to consider separately finding the solution $p_k(\mathbf{h})$ that satisfies the k th KKT condition in (4) for each user k , while keeping the powers of all other users $j \neq k$ as fixed and known quantities.

Let us concentrate on user k , and fix $p_j(\mathbf{h})$, $j \neq k$. It can be shown that, the solution to (6) subject to the average power constraint on $p_k(\mathbf{h})$ should satisfy the KKT condition for the single user problem,

$$\sum_{i=1}^k \frac{\mu_i - \mu_{i-1}}{a_{ki}(\mathbf{h}) + p_k(\mathbf{h})} \leq \tilde{\lambda}_k, \quad \forall \mathbf{h} \quad (7)$$

A side remark here: $\tilde{\lambda}_k$ is in general different from the Lagrange multiplier λ_k in (4), since the powers we have fixed for the

other users need not be the optimal powers. Eventually, since the iterative algorithm converges to the optimal powers, we know that $\tilde{\lambda}_k$ will converge to λ_k .

We will next argue how this condition can be interpreted as a ‘‘generalized’’ waterfilling. First assume no power has yet been allocated to any channel state. Define the inverse of the left hand side of (7) evaluated at $p_k(\mathbf{h}) = 0$ for all \mathbf{h} by,

$$b_k(\mathbf{h}) = \left(\sum_{i=1}^k \frac{\mu_i - \mu_{i-1}}{a_{ki}(\mathbf{h})} \right)^{-1} \quad (8)$$

Then, sort $b_k(\mathbf{h})$ over all channel states \mathbf{h} in increasing order. Since user k has to satisfy its average power constraint, it has to put some power to a non-zero probability subset, say Ω , of all possible channel states. At the channel states where user k transmits with positive power, (7) needs to be satisfied with equality. Let user k start pouring some of its available power to the state which gives the lowest $b_k(\mathbf{h})$, say \mathbf{h}' . Next, pick another state \mathbf{h}'' , such that $b_k(\mathbf{h}') < b_k(\mathbf{h}'')$. User k starts transmitting at \mathbf{h}'' only if (i) it has already poured some powers $q_k(\mathbf{h})$ to all states \mathbf{h} such that $b_k(\mathbf{h}) < b_k(\mathbf{h}'')$, (ii) it still has some power left to allocate, and (iii) the already allocated powers satisfy

$$\sum_{i=1}^k \frac{\mu_i - \mu_{i-1}}{a_{ki}(\mathbf{h}) + q_k(\mathbf{h})} = b_k^{-1}(\mathbf{h}''), \quad \forall \mathbf{h} : b_k(\mathbf{h}) \leq b_k(\mathbf{h}'') \quad (9)$$

Before going any further, using the current construction, let us revisit the sum capacity case in [6], [7] where μ_i , $i = 1, \dots, K$, are all equal to 1. In this case, from (8), $b_k(\mathbf{h}) = a_{k1}(\mathbf{h})$, and it can be easily seen that the described procedure produces the ordinary waterfilling solution; user k will pour its power over $a_{k1}(\mathbf{h}) = b_k(\mathbf{h})$, until all the available power is used. The optimal power value at \mathbf{h} is the difference between the water level $1/\tilde{\lambda}_k$ and the base level $b_k(\mathbf{h})$, whenever the difference is positive; it is zero otherwise, i.e.,

$$p_k(\mathbf{h}) = \left(\frac{1}{\tilde{\lambda}_k} - b_k(\mathbf{h}) \right)^+ \quad (10)$$

The main subtlety in solving for the optimal powers in the arbitrary μ_i s case is that, there are more than one terms that involve $p_k(\mathbf{h})$ on the left hand side of (7), and thus the optimal $p_k(\mathbf{h})$ is no longer given by a nice expression such as (10), but is rather the solution to a polynomial equation. Therefore, the optimal power levels lose their waterfilling interpretation. However, we can still see the procedure described here as a type of waterfilling, as it gradually equalizes the base levels $b_k(\mathbf{h})$, and solves for the power levels required for such equalization, hence the name ‘‘generalized’’ waterfilling.

Generalized waterfilling yields the optimum power allocation because of the fact that by construction, the KKT conditions are satisfied when all average power is used. To see this, let us denote the left hand side of (9) by $L(\mathbf{h}, q_k(\mathbf{h}))$. We keep increasing $b_k(\mathbf{h}'')$ on the right hand side of (9) gradually. Letting $p_k(\mathbf{h}) = q_k(\mathbf{h})$ when the solution $q_k(\mathbf{h})$ obtained from (9) satisfies the average power constraints, and taking $\tilde{\lambda}_k \triangleq L(\mathbf{h}, p_k(\mathbf{h}))$, we see

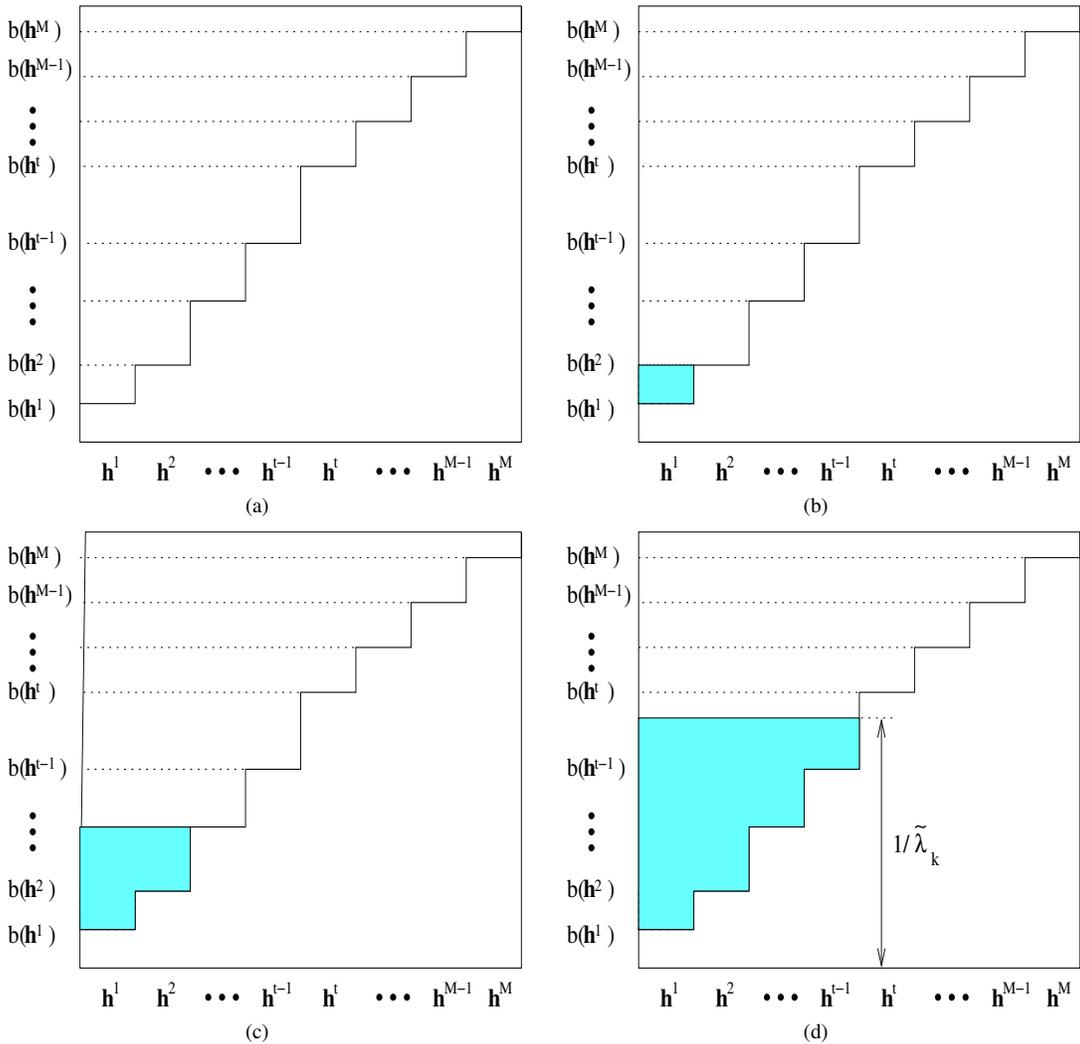


Fig. 2. Illustration of the generalized waterfilling.

that the solution $p_k(\mathbf{h})$ satisfies the KKT conditions and it is optimal.

In order to better visualize how the generalized waterfilling is performed, we consider a simple example with $K = 2$ and with discrete joint channel states \mathbf{h}^i , $i = 1, \dots, M$. Without loss of generality, let us assume $b_k(\mathbf{h}^1) < \dots < b_k(\mathbf{h}^M)$. Figure 2 shows the generalized waterfilling procedure. The ordered values $b_k(\mathbf{h}^i)$ are illustrated in Figure 2(a). First, using (9), we solve for the amount of power $q_k(\mathbf{h}^1)$ that will level $L(\mathbf{h}^1, q_k(\mathbf{h}^1))$ and $b_k(\mathbf{h}^2)$, so that the water level is $b_k(\mathbf{h}^2)$, as shown in Figure 2(b). It can be easily shown that $q_k(\mathbf{h}^1)$ is the only non-negative solution to a k^{th} order polynomial equation, obtained from (9). In this particular example, the available average power is not yet completely used in this first step, so we repeat the same procedure at both \mathbf{h}^1 and \mathbf{h}^2 , i.e., we solve for $q_k(\mathbf{h}^1)$ and $q_k(\mathbf{h}^2)$ that will level $L(\mathbf{h}^1, q_k(\mathbf{h}^1))$, $L(\mathbf{h}^2, q_k(\mathbf{h}^2))$ and $b_k(\mathbf{h}^3)$ (see Figure 2(c)). We continue this procedure until we see that although the water levels can be made equal at $b_k(\mathbf{h}^{t-1})$ while satisfying the average power constraint, it is not possible to equalize the water levels

at $b_k(\mathbf{h}^t)$, since the available average power falls short of the required average power that is needed for such equalization. At this point, we know that the final water level, i.e., the true value of $1/\tilde{\lambda}_k$ that will satisfy the KKT conditions together with $q_k(\mathbf{h}^i)$ obtained from (9) should lie between $b_k(\mathbf{h}^{t-1})$ and $b_k(\mathbf{h}^t)$, and we can find it by searching between these two values until the $q_k(\mathbf{h}^i)$, $i = 1, \dots, t-1$, satisfy the average power constraint with equality. Figure 2(d) illustrates this last step, and the final value of $\tilde{\lambda}_k$ that satisfies the KKT conditions.

Note that, by letting $\mu_1 = \dots = \mu_K$, we recover the traditional waterfilling solution in [6], [7], since only the first term survives in the KKT conditions. On the other hand, if we let $s_i = 1$ for $i = 1, \dots, K$, the generalized waterfilling algorithm solves the resource allocation problem in [3] for scalar MAC.

IV. SIMULATION RESULTS

In this section, we present some simulation results for the generalized iterative waterfilling algorithm. In our simulations, we pick the number of users $K = 2$, so that our results such as the capacity regions and the optimum power allocations can be

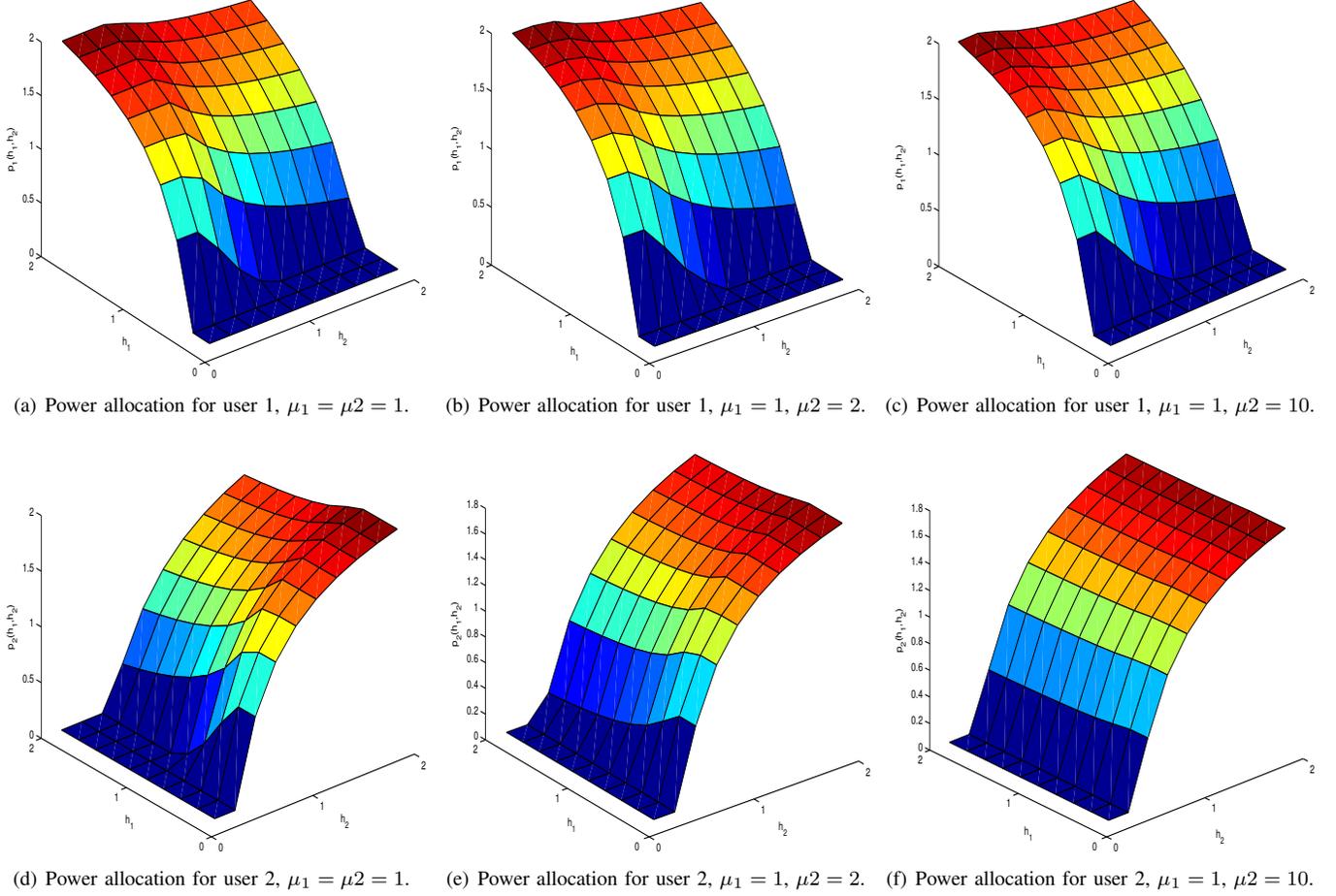


Fig. 3. Power distributions for different values of priorities.

easily visualized. The channel states h_1 and h_2 are chosen to be independent uniform random variables, each taking values from the discrete set $\{0.2:0.2:2\}$. The processing gain is chosen to be $N = 2$, the noise variance is $\sigma^2 = 1$, and both users have an average power constraint equal to 1.

First, in order to observe the effect of the priorities μ_i on the optimum power allocation, we plot the optimum power allocation policies for both users for two different sets of (μ_1, μ_2) values. We fix the signature sequences of the users to be $\mathbf{s}_1 = [1/\sqrt{2} \ 1/\sqrt{2}]^T$, and $\mathbf{s}_2 = [1 \ 0]^T$. Figures 3(a) and 3(d) correspond to the sum capacity maximizing power control policies, i.e., to $(\mu_1, \mu_2) = (1, 1)$. In each figure, the height of the surface corresponds to the power allocated to each channel state. We see that the two users perform simultaneous waterfilling, which was also observed in [6], [7]. Here, each of the users tend to transmit with less power over the channel states where the other user is stronger, and due to the symmetry of the problem, the power allocation policies are symmetric. When we choose unequal priorities $(\mu_1, \mu_2) = (1, 2)$, we observe in Figures 3(b) and 3(e) that the power allocation for user 1 does not change significantly, but user 2 pours more power to channel states where it transmitted with considerably less power in the symmetric

priorities case. If we increase μ_2 even further, and solve for the case when $(\mu_1, \mu_2) = (1, 10)$, we see in Figure 3(f) that the power allocation policy of user 2 converges nearly to single user waterfilling. Since the priority of user 1 is much less than that of user 2, user 2, while trying to maximize the weighted sum of rates, acts as if it is alone in the system in allocating its power. The power allocation of user 1, given in Figure 3(c), is not significantly different from the previous two cases.

In Figure 4, we give the capacity region of the fading CDMA channel, for different values of correlations between the signature sequences. The regions are formed by finding the optimal power allocation policies for a large set of (μ_1, μ_2) values, and then by using these allocation policies to compute the corresponding (R_1, R_2) pairs. The case when the correlation is $\rho = 1$ corresponds to the identical signature sequences case, in which case the boundary of the capacity region is strictly convex, and each point on the surface can be achieved by a power control policy, without timesharing. Note that, this setting also covers the scalar MAC case in [3], and the properties derived in [3] for the capacity region are observed here. When we decrease the correlation between the sequences, we begin to observe a flat portion on the capacity region, which agrees with the findings

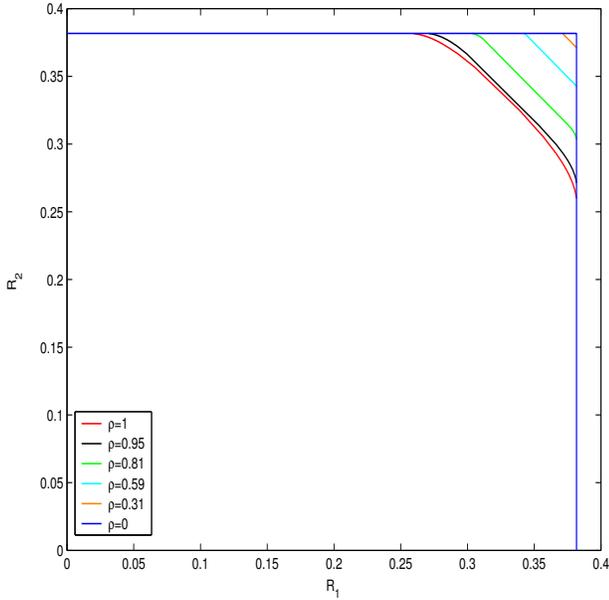


Fig. 4. Capacity region of a two user fading CDMA channel for several correlation values among the signature sequences.

of [6]. As we further decrease the correlation, eventually the sequences become orthogonal and the capacity region becomes the rectangular region whose boundaries are single user limits, as expected.

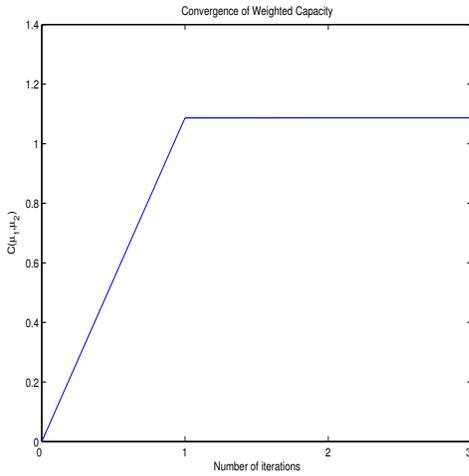


Fig. 5. Convergence of generalized iterative waterfilling.

Finally, in Figure 5, we show an example of the convergence of the generalized iterative waterfilling algorithm for the simple

system considered here; the powers converge after only three iterations, and the optimum weighted capacity value is almost attained after one round of iterations.

V. CONCLUSIONS

We have characterized the power allocation policies that achieve arbitrary rate tuples on the boundary of the capacity region of a fading CDMA channel. The optimal power allocation policy for a given set of priorities μ_i is the joint solution to the extended KKT conditions for all users. Since the KKT inequalities appear difficult to solve analytically, we have provided a one-user-at-a-time iterative power allocation algorithm that converges to the optimum solution. We showed that, each iteration of this algorithm corresponds to solving for the power levels of the user of interest at all fading states, so that the power allocation satisfies the single-user KKT conditions. We have also provided a “generalized” waterfilling interpretation for the power allocation procedure as it operates by gradually equating the levels of “water” poured on top of certain base levels, which are functions of the channel states, power levels of other users, and the priorities μ_i .

REFERENCES

- [1] A. J. Goldsmith and P. P. Varaiya. Capacity of fading channels with channel side information. *IEEE Transactions on Information Theory*, 43(6):1986–1992, November 1997.
- [2] R. Knopp and P. A. Humblet. Information capacity and power control in single-cell multiuser communications. In *IEEE International Conference on Communications*, June 1995.
- [3] S. Hanly and D.N.C. Tse. Multiaccess fading channels-part I: Polymatroid structure, optimal resource allocation and throughput capacities. *IEEE Transactions on Information Theory*, 44(7):2796–2815, November 1998.
- [4] W. Yu, W. Rhee, S. Boyd, and J. M. Cioffi. Iterative water-filling for Gaussian vector multiple-access channels. *IEEE Transactions on Information Theory*, 50(1):145–152, January 2004.
- [5] W. Yu, W. Rhee, and J. M. Cioffi. Optimal power control in multiple access fading channels with multiple antennas. In *IEEE International Conference on Communications*, pages 575–579, June 2001.
- [6] O. Kaya and S. Ulukus. Optimum power control for CDMA with deterministic sequences in fading channels, *IEEE Transactions on Information Theory*, submitted December 2002, revised December 2003.
- [7] O. Kaya and S. Ulukus. Optimum power control for fading CDMA with deterministic sequences. In *40th Annual Allerton Conference on Communications, Control and Computing, Monticello, IL, October 2002*.
- [8] S. Verdú. *Multiuser Detection*. Cambridge University Press, 1998.
- [9] E. Biglieri, J. Proakis, and S. Shamai (Shitz). Fading channels: Information-theoretic and communications aspects. *IEEE Transactions on Information Theory*, 44(6):2619–2692, October 1998.