

ON Sensitivity in
Semistate Described
Linear Systems

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Semistate Equations in Canonical Form:

$$a \dot{x} + B(x,t) = D u$$

$$y = F x$$

- $x =$ semistate, k -vector
 - $u =$ input, n -vector
 - $y =$ output, m -vector
- $\dot{} = d/dt$

a, D, F constant & generally singular.

Linear, time-invariant

$$B(x,t) = B \cdot x$$

with B constant

Variation with a parameter p

$$A(p)\dot{x} + B(p) \cdot x = D(p) \cdot u$$

$$y = F(p) \cdot x$$

assume u is independent of p ,
then

$$x = x(t, p)$$

$$y = y(t, p)$$

p may be a system element,
such as an amplifier gain or
resistor resistance, or it may
be something like temperature
or radiation that affects system
elements.

Sensitivity

In sensitivity analysis we are interested in the effects of changes in the parameter ρ .

Time Domain

$$\frac{\partial y(t, \rho)}{\partial \rho}$$

(a function of t)

Frequency Domain

scalar transfer function sensitivity

$$T(s, \rho) = Y(s, \rho) / U(s)$$

$$S_{\rho}^T = \frac{\rho}{T} \frac{\partial T}{\partial \rho}$$

(a function of s
giving % change.)

Time Domain Sensitivity

Given

$$a \dot{x} + B \cdot x = D u$$

$$y = \mathcal{H} x$$

we wish to find $y' = \frac{\partial y}{\partial p}$

$$a' \dot{x} + a \dot{x}' + B' \cdot x + B \cdot x' = D' u$$

$$y' = \mathcal{H}' x + \mathcal{H} \cdot x'$$

Therefore simply resolve the original-like equations

$$a(x') + B(x') = D' u - \underbrace{a' \dot{x} + B' x}_{\text{known from solving original equations}}$$

$$[x'(0) \equiv 0 \text{ as } x(0, p) = x(0)]$$

$$y' = \mathcal{H}' x' + \mathcal{H}' x$$

Note that this process would be "simplified" if a, D, \mathcal{H} were independent of p .

Transformation to Simplified Form

Given

$$A \dot{x} + Bx = Du$$

$$y = Mx$$

Find nonsingular P, Q for

$$PAQ = I_r + O_{k-r}$$

$\dot{+}$ = direct sum

$I_r = r \times r$ identity

$O_{k-r} = (k-r) \times (k-r)$
zero

Let

$$x = Q \hat{x}, \quad \hat{x} = \begin{bmatrix} \hat{x}_1 \\ \hat{x}_2 \end{bmatrix}$$

$$\hat{B} = P B Q = \begin{bmatrix} \hat{B}_{11} & \hat{B}_{12} \\ \hat{B}_{21} & \hat{B}_{22} \end{bmatrix}$$

$$\hat{D} = P D = \begin{bmatrix} \hat{D}_1 \\ \hat{D}_2 \end{bmatrix}$$

$$\hat{M} = M Q = [\hat{M}_1, \hat{M}_2]$$

Giving

$$\begin{bmatrix} I_r & 0 \end{bmatrix} \dot{\hat{x}} + \hat{B} \hat{x} = \hat{D} u$$

$$y = \hat{M} \hat{x}$$

Let

$$\underline{x}_3 = u$$

$$\underline{x}_4 = \hat{\mathcal{H}} \hat{x}$$

$$\underline{x}_1 = \hat{x}_1$$

$$\underline{x}_2 = \hat{x}_2$$

and $\underline{x} = \begin{bmatrix} \underline{x}_1 \\ \underline{x}_2 \\ \underline{x}_3 \\ \underline{x}_4 \end{bmatrix}$

Then

$$\begin{bmatrix} 1_n & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \\ 0 & 0 & 0 & 0 \end{bmatrix} \dot{\underline{x}} + \begin{bmatrix} \hat{\mathcal{B}}_{11} & \hat{\mathcal{B}}_{12} & -\hat{\mathcal{D}}_1 & 0 \\ \hat{\mathcal{B}}_{21} & \hat{\mathcal{B}}_{22} & -\hat{\mathcal{D}}_2 & 0 \\ 0 & 0 & 1_m & 0 \\ -\hat{\mathcal{H}}_1 & -\hat{\mathcal{H}}_2 & 0 & 1_m \end{bmatrix} \underline{x} = \begin{bmatrix} 0 \\ 0 \\ 1_m \\ 0 \end{bmatrix} u$$

$$y = [0, 0, 0, 1_m] \underline{x}$$

This has all its dependence in \underline{B} , hence

$$\underline{A} \dot{\underline{x}} + \underline{B} \underline{x} = \underline{D} u = \underline{f}(t)$$

$$y = \underline{H} \underline{x}$$

gives

$$\underline{A} (\dot{\underline{x}}) + \underline{B} (\underline{x}) = -\underline{B}' \cdot \underline{x} = \underline{f}(t)$$

$$\dot{\underline{x}} = \underline{H} \cdot \underline{x}'$$

[note $\underline{x}'(0, \rho) \neq 0$ though, unless $\underline{x}(0) = 0$]

Therefore can solve via the
Drazin inverse

$$Ax + B(x_0)x = Du$$

$$y = \mathcal{H}x$$

Then resolve

$$Ax' + B(x_0)x' = -B'x$$

$$y' = \mathcal{H}x'$$

to get y'

Drazin inverse solution

$$\begin{aligned}
 \dot{x} + Bx &= f(t) \\
 y &= Fx
 \end{aligned}$$

has the solution

$$\begin{aligned}
 x(t) &= \tilde{a} \tilde{a}^D e^{-\tilde{a}^D \tilde{B} t} x_0 + \tilde{B}^D (1 - \tilde{a} \tilde{a}^D) \sum_{i=0}^{I-1} (-1)^i (\tilde{a} \tilde{B}^D)^i \tilde{f}^{(i)}(t) \\
 &+ e^{-\tilde{a}^D \tilde{B} t} 1(t) * (\tilde{a}^D \tilde{f}(t))
 \end{aligned}$$

where

$$\tilde{a} = (a + B)^{-1} a, \quad \tilde{B} = (a + B)^{-1} B, \quad \tilde{f} = (a + B)^{-1} f$$

and $D = \text{Drazin inverse}$, $I = \text{index of } a$
 $*$ = convolution $1(\cdot) = \text{unit step}$

If $A = T \begin{bmatrix} C & 0 \\ 0 & N \end{bmatrix} T^{-1}$, $C = \text{nonsingular part}$
 $N = \text{nilpotent part}$

then $A^D = T \begin{bmatrix} C^{-1} & 0 \\ 0 & 0 \end{bmatrix} T^{-1}$; index = smallest I for $N^I = 0$

Note: can form ' of $x(t) = x(t, \infty)$ from above expression = big mess.

Transfer Function

Scalar Case

Put p in \mathcal{B}

$$ax + \mathcal{B}(p)x = Du$$

$$y = \mathcal{H}x$$

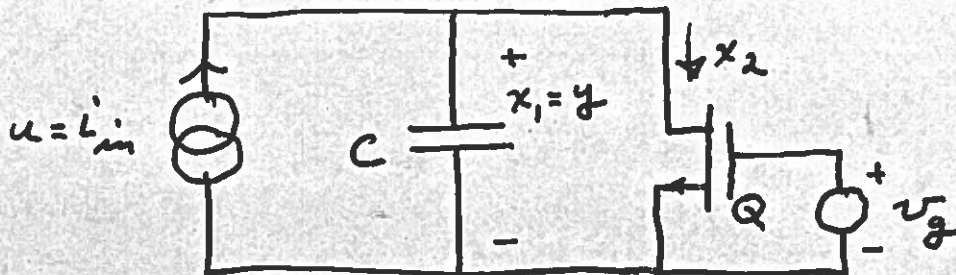
$$Y = T(a, p)U$$

$$T(a, p) = \mathcal{H}(a\mathcal{A} + \mathcal{B})^{-1}\mathcal{D}$$

$$\frac{\partial T}{\partial p} = \mathcal{H}(a\mathcal{A} + \mathcal{B})^{-1}(-\mathcal{B}')\mathcal{D}$$

$$S_p^T = \frac{p \{ \mathcal{H}(a\mathcal{A} + \mathcal{B})^{-1}(-\mathcal{B}')\mathcal{D} \}}{\mathcal{H}(a\mathcal{A} + \mathcal{B})^{-1}\mathcal{D}}$$

Example:



$$C \dot{x}_1 + x_2 = u$$

$$x_2 - \beta(v_g - V_T)x_1 = 0 \quad \left. \begin{array}{l} \text{assume } Q \text{ in} \\ \text{resistive} \\ \text{region} \end{array} \right\}$$

$$y = x_1$$

or

$$\begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} \dot{x} + \begin{bmatrix} 0 & 1/C \\ -\beta(v_g - V_T) & 1 \end{bmatrix} x = \begin{bmatrix} 1/C \\ 0 \end{bmatrix} u$$

$$y = [1, 0] x$$

Let $p = v_g$.

Therefore are interested in the output voltage sensitivity with respect to the gate voltage (whose variation varies the transistor resistance)

Now normalize:

multiply 2nd row by $\frac{1}{\beta V_T}$ and then

let $\underline{x}_2 = \frac{1}{\beta V_T} x_2$. Multiply 1st row by

$\frac{c}{\beta V_T}$. Let $\underline{u} = \frac{1}{\beta V_T} u$. Change time scale

by $\tau = \left(\frac{\beta V_T}{c}\right) t$ and change v_g scale by

$V_g = v_g / V_T$. Let $\underline{x}_1(\tau) = x_1(t)$, $\underline{x}_2(\tau) = \underline{x}_2(t)$,

$\underline{u}(\tau) = \underline{u}(t)$. Then replace all variables

by their original symbols to get

$$\begin{bmatrix} 1 & 0 \\ 0 & 0 \end{bmatrix} \dot{x} + \begin{bmatrix} 0 & 1 \\ -(V_g - 1) & 1 \end{bmatrix} x = \begin{bmatrix} 1 \\ 0 \end{bmatrix} u$$

$$y = [1, 0] x$$

Then

$$(aA + B)^{-1} = \frac{1}{a + (V_g - 1)} \begin{bmatrix} 1 & -1 \\ (V_g - 1) & a \end{bmatrix}$$

$$\tilde{a} = (aA + B)^{-1} a = \frac{1}{a + (V_g - 1)} \begin{bmatrix} 1 & 0 \\ (V_g - 1) & 0 \end{bmatrix}$$

$$\tilde{B} = \frac{1}{a + (V_g - 1)} \begin{bmatrix} V_g - 1 & 0 \\ -a(V_g - 1) & a + (V_g - 1) \end{bmatrix}$$

$$\tilde{D} = \frac{1}{a + (V_g - 1)} \begin{bmatrix} 1 \\ (V_g - 1) \end{bmatrix} = (aA + B)^{-1} D$$

Now

$$\tilde{\alpha} = \frac{1}{a+(v_g-1)} \begin{bmatrix} 1 & 0 \\ v_g-1 & 0 \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ v_g-1 & 1 \end{bmatrix} \begin{bmatrix} \frac{1}{a+(v_g-1)} & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ -(v_g-1) & 1 \end{bmatrix}$$

$$= T \begin{bmatrix} C & 0 \\ 0 & N \end{bmatrix} T^{-1} \quad ; I = 1$$

$$\text{or } \tilde{\alpha}^D = \begin{bmatrix} 1 & 0 \\ v_g-1 & 1 \end{bmatrix} \begin{bmatrix} a+(v_g-1) & 0 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} 1 & 0 \\ -(v_g-1) & 1 \end{bmatrix}$$

$$= \begin{bmatrix} a+(v_g-1) & 0 \\ (v_g-1)(a+(v_g-1)) & 0 \end{bmatrix}$$

$$\tilde{\beta}^D = \tilde{\beta}^{-1} = \begin{bmatrix} \frac{a+(v_g-1)}{(v_g-1)} & 0 \\ -a & 1 \end{bmatrix}$$

$$\tilde{\alpha} \tilde{\alpha}^D = \begin{bmatrix} 1 & 0 \\ v_g-1 & 0 \end{bmatrix} \quad \tilde{\alpha}^D \tilde{\beta} = \begin{bmatrix} v_g-1 & 0 \\ (v_g-1)^2 & 0 \end{bmatrix}$$

$$\tilde{\beta}^D [I_n - \tilde{\alpha} \tilde{\alpha}^D] = \begin{bmatrix} 0 & 0 \\ -(v_g-1) & 1 \end{bmatrix}, \quad \tilde{\alpha} \tilde{\beta}^D = \begin{bmatrix} \frac{1}{v_g-1} & 0 \\ 1 & 0 \end{bmatrix}$$

$$\tilde{\alpha}^D \tilde{\beta} = \begin{bmatrix} 1 \\ v_g-1 \end{bmatrix}$$

$$x_0 = \begin{bmatrix} x_{01} \\ x_{02} \end{bmatrix} \text{ and here } x_0' = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

$$\tilde{\beta}^D [I_n - \tilde{\alpha} \tilde{\alpha}^D] x_0 = \begin{bmatrix} 0 & 0 \\ -(v_g-1) & 1 \end{bmatrix} \cdot \frac{1}{a+(v_g-1)} \begin{bmatrix} 1 \\ v_g-1 \end{bmatrix} = \begin{bmatrix} 0 \\ 0 \end{bmatrix}$$

$$x(t, v_g) = \begin{bmatrix} 1 & 0 \\ v_g - 1 & 0 \end{bmatrix} e^{-\begin{bmatrix} v_g - 1 & 0 \\ (v_g - 1)^2 & 0 \end{bmatrix} t} x_0 + e^{-\begin{bmatrix} v_g - 1 & 0 \\ (v_g - 1)^2 & 0 \end{bmatrix} t} 1(t) * \begin{bmatrix} 1 \\ v_g - 1 \end{bmatrix} u(t)$$

Bsp

$$e^{-\begin{bmatrix} v_g - 1 & 0 \\ (v_g - 1)^2 & 0 \end{bmatrix} t} = \begin{bmatrix} e^{-(v_g - 1)t} & 0 \\ (v_g - 1)(e^{-(v_g - 1)t} - 1) & 1 \end{bmatrix}$$

or

$$x(t, v_g) = \begin{bmatrix} e^{-(v_g - 1)t} & 0 \\ (v_g - 1)e^{-(v_g - 1)t} & 0 \end{bmatrix} \begin{bmatrix} x_{0,1} \\ x_{0,2} \end{bmatrix} +$$

$$+ \begin{bmatrix} e^{-(v_g - 1)t} \\ (v_g - 1)e^{-(v_g - 1)t} \end{bmatrix} 1(t) * u(t)$$

$$= \begin{bmatrix} 1 \\ (v_g - 1) \end{bmatrix} \left\{ e^{-(v_g - 1)t} x_{0,1} + e^{-(v_g - 1)t} 1(t) * u(t) \right\}$$

To get $x'(t, v)$ we repeat with Du replaced by $-B'x$. Here

$$B = \begin{bmatrix} 0 & 1 \\ -(v_g-1) & 1 \end{bmatrix} \text{ or } B' = \begin{bmatrix} 0 & 0 \\ -1 & 0 \end{bmatrix} \text{ and}$$

$$-B'x(t, v_g) = \begin{bmatrix} 0 \\ 1 \end{bmatrix} \left\{ e^{-(v_g-1)t} x_0 + e^{-(v_g-1)t} 1(t) * u(t) \right\}$$

and

$$\tilde{f} \rightarrow -(A+B)B'x$$

$$= \frac{1}{A+(v_g-1)} \begin{bmatrix} -1 \\ A \end{bmatrix} \left\{ e^{-(v_g-1)t} x_0 + e^{-(v_g-1)t} 1(t) * u(t) \right\}$$

$$\tilde{A} \tilde{f} = \begin{bmatrix} -1 \\ -(v_g-1) \end{bmatrix} \left\{ e^{-(v_g-1)t} x_0 + e^{-(v_g-1)t} 1(t) * u(t) \right\}$$

and

note $\neq 0$ no jump in x' at $t=0+$

$$x'(t, v_g) = \begin{bmatrix} 0 \\ 0 \end{bmatrix} + \begin{bmatrix} 0 & 0 \\ -(v_g-1) & 1 \end{bmatrix} \frac{1}{A+(v_g-1)} \begin{bmatrix} -1 \\ A \end{bmatrix} \left\{ e^{-(v_g-1)t} x_0 + e^{-(v_g-1)t} 1(t) * u(t) \right\}$$

$$+ \begin{bmatrix} e^{-(v_g-1)t} & 0 \\ (v_g-1)e^{-(v_g-1)t} & 1 \end{bmatrix} 1(t) * \begin{bmatrix} -1 \\ -(v_g-1) \end{bmatrix} \left\{ e^{-(v_g-1)t} x_0 + e^{-(v_g-1)t} 1(t) * u(t) \right\}$$

as $t > 0$
has a $1(t)$ multiplier

$$= \begin{bmatrix} 0 \\ 1 \end{bmatrix} \left\{ e^{-(v_g-1)t} x_0 + e^{-(v_g-1)t} 1(t) * u(t) \right\}$$

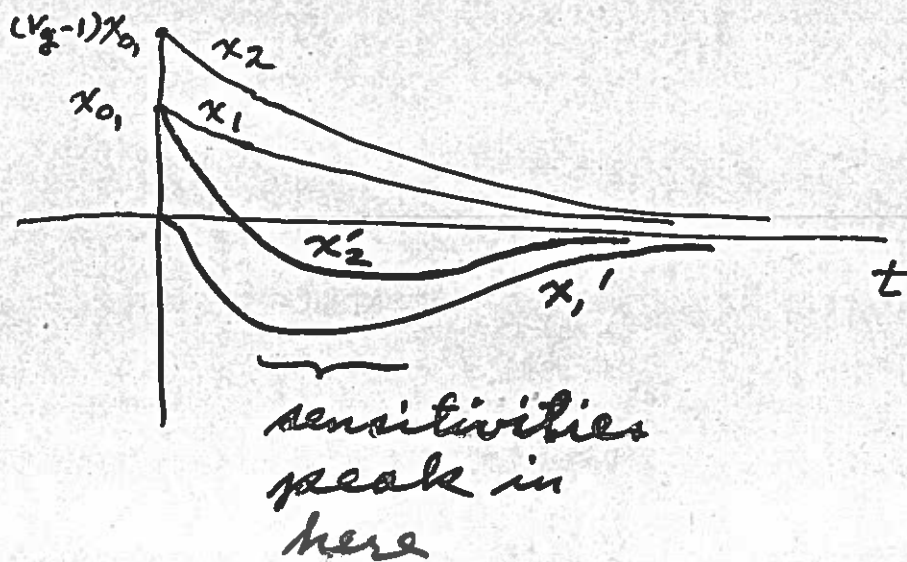
$$- \begin{bmatrix} e^{-(v_g-1)t} \\ (v_g-1)e^{-(v_g-1)t} - 1 \end{bmatrix} 1(t) * \left\{ e^{-(v_g-1)t} x_0 1(t) + e^{-(v_g-1)t} 1(t) * u(t) \right\}$$

Special case - initial conditions only

$$u(t) \equiv 0$$

$$x'(t, v_g) = x_0 \left\{ \begin{bmatrix} 0 \\ 1 \end{bmatrix} e^{-(v_g-1)t} - \begin{bmatrix} e^{-(v_g-1)t} \\ (v_g-1)(e^{-(v_g-1)t} - 1) \end{bmatrix} e^{-(v_g-1)t} \right\}$$

with $x(t) = \begin{bmatrix} 1 \\ v_g-1 \end{bmatrix} e^{-(v_g-1)t} x_0$



Example Transfer Function Sensitivity

$$S_{V_g}^T = \frac{V_g \{ \mathcal{H}(a s + B)^{-1} (-B') (a s + B)^{-1} \mathcal{D} \}}{\mathcal{H}(a s + B)^{-1} \mathcal{D}}$$

$$(a s + B)^{-1} = \frac{1}{a + (V_g - 1)} \begin{bmatrix} 1 & -1 \\ V_g - 1 & a \end{bmatrix}$$

$$B' = \begin{bmatrix} 0 & 0 \\ -1 & 0 \end{bmatrix}, \quad \mathcal{H} = [1, 0], \quad \mathcal{D} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$\begin{aligned} T(s, r) &= \mathcal{H}(a s + B)^{-1} \mathcal{D} = [1, 0] \left\{ \frac{1}{a + (V_g - 1)} \begin{bmatrix} 1 & -1 \\ V_g - 1 & a \end{bmatrix} \right\} \begin{bmatrix} 1 \\ 0 \end{bmatrix} \\ &= \frac{1}{a + (V_g - 1)} \end{aligned}$$

and

$$\mathcal{H}(a s + B)^{-1} = \frac{1}{a + (V_g - 1)} [1 \quad -1]$$

$$-B'(a s + B)^{-1} \mathcal{D} = \begin{bmatrix} 0 & 0 \\ -1 & 0 \end{bmatrix} \frac{1}{a + (V_g - 1)} \begin{bmatrix} 1 \\ V_g - 1 \end{bmatrix} = \frac{1}{a + (V_g - 1)} \begin{bmatrix} 0 \\ 1 \end{bmatrix}$$

$$\text{or } \mathcal{H}(a s + B)^{-1} (-B') (a s + B)^{-1} \mathcal{D} = \frac{-1}{(a + (V_g - 1))^2}$$

$$S_{V_g}^T = \frac{+ V_g \left(\frac{-1}{(a + (V_g - 1))^2} \right)}{\frac{1}{a + (V_g - 1)}} = \boxed{\frac{-V_g}{a + (V_g - 1)}}$$

which checks $S_{V_g}^T = \frac{r}{T} \cdot \frac{\partial T}{\partial r}$

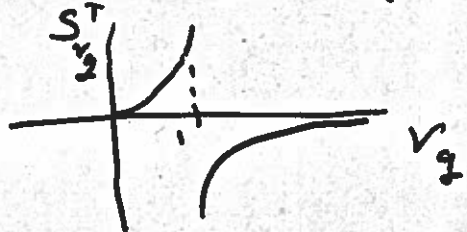
$$T = \frac{V_c}{I_{in}} = \text{input impedance}$$

$$= \frac{1}{A + (V_g - 1)}$$

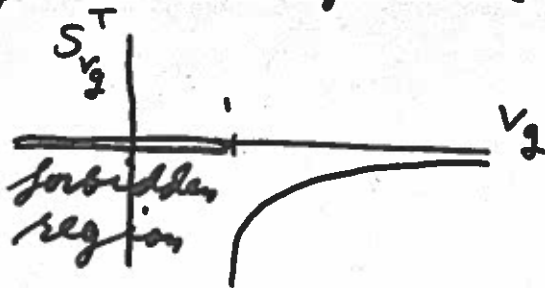
$$S_p^T = S_{V_g}^{V_c/I_{in}} = \frac{-V_g}{A + (V_g - 1)}$$

S_p^T decreases in magnitude as frequency $\omega = s/j$ increases. at

DC, $\omega = 0$, $S_p^T = \frac{-V_g}{V_g - 1}$



Useful for tuning near $V_g = 1$, but by assumption $V_g = v_g/V_T > 1$ or



Conclusion

The semi-state equations can be directly used for sensitivity calculations.

The calculations are somewhat messy but could be well-organized for use in design from semi-state specifications.