

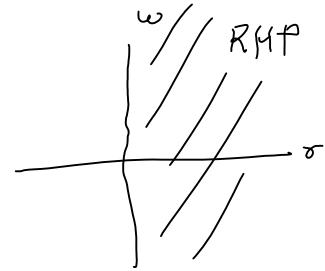
$$\begin{aligned} \operatorname{Re} V^{T*} I &= \frac{1}{2} (V^{T*} I + V^T I^*) = \frac{1}{2} (V^{T*} I + I^{T*} V) \quad , \quad I = Y(a) V \\ &= \frac{1}{2} (V^{T*} [Y(a) V] + [V^{T*} Y^{T*}(a)] V) \quad \text{if } Y \text{ is } n \times n, V \text{ is an } n\text{-vector} \\ &= V^{T*} \left[ \frac{Y(a) + Y^{T*}(a)}{2} \right] V \geq 0 \text{ on } a = j\omega, \\ &\quad \Rightarrow \text{Passive} \end{aligned}$$

$Y^H(a) = \frac{1}{2} (Y(a) + Y^{T*}(a))$  is positive semidefinite for passive circuits on  $a = j\omega$

If passive no poles in  $\sigma > 0$  if rational (causal stable)

$Y^*(a) = Y(a^*)$

Look at  $e^{-V^{T*} Y(a) V}$  is analytic in  $\sigma > 0$



by the maximum modulus theorem

$|e^{-V^{T*} Y(a) V}|$  is maximum on  $\sigma = 0$   
 $|e^{-\operatorname{Re} V^{T*} Y V - j \operatorname{Im} V^{T*} Y V}| = e^{-\operatorname{Re} V^{T*} Y(a) V}$  is max of RHP on  $\sigma = 0$

$\Rightarrow \operatorname{Re} V^{T*} Y(a) V \geq 0$  in  $\sigma > 0 \Rightarrow \frac{Y(a) + Y^{T*}(a)}{2}$  is positive semidefinite in  $\sigma > 0$

Positive real conditions are

- 1)  $Y^*(a) = Y(a^*)$  for  $\sigma > 0$  real circuit
- 2)  $Y(a)$  is analytic for  $\sigma > 0$  stable
- 3)  $Y(a) + Y^{T*}(a)$  is positive semi-definite in  $\sigma > 0$ ; passive

if rational then call BR  $\iff$  can build a finite RLC gyrator circuit

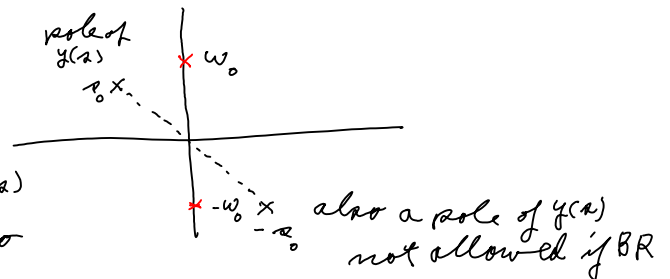
$\Rightarrow$  BR = physically realizable

Lossless if  $Y(a) + Y(-a) = 0_n \Rightarrow Y(a) = -Y(-a)$

for  $n=1, y(a) = -y(-a)$

$\therefore$  for lossless  $y(a)$ , BR, all poles are on  $j\omega$  axis

also holds for LBR: lossless BR,  $z(a)$   
 $\Rightarrow$  implies zeroes of  $y(a)$  are also on  $j\omega$  axis



near a pole  $y(a) \approx \frac{k}{(a - j\omega_0)^n} + \dots = \frac{k = |k| e^{j\theta}}{R^n e^{j\theta} e^{jn\phi}}$

$y(a) = \left(\frac{|k|}{R^n}\right) e^{j(a\tau - \theta^n)}$ ,  $-\pi/2 < \theta < \pi/2$



will mean  $k \neq 0, n = 1$

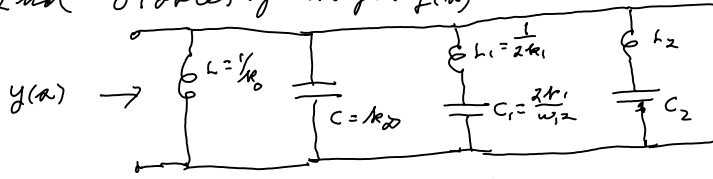
$\operatorname{Re} y(a) = \left(\frac{|k|}{R^n}\right) \cos(\omega\tau - \theta^n) \Rightarrow \omega\tau = 0, n = 1$   
 in semicircle

\$\Rightarrow\$ any finite pole is simple (\$N=1\$) & with real positive residue \$k\$

$$y(s) = \sum_{\substack{\text{poles} \\ \omega_j}} \frac{k_j}{s - \omega_j} + k_\infty s, \quad k_j > 0, k_\infty > 0$$

$$= \frac{k_0}{s} + k_\infty s + \underbrace{\frac{k_1}{s - \omega_1} + \frac{k_1}{s + \omega_1}}_{\frac{2k_1 s}{s^2 + \omega_1^2}} + \frac{k_2}{s - \omega_2} + \frac{k_2}{s + \omega_2} + \dots$$

2nd Foster form for \$y(s)\$



$$y(s) = \frac{k_0}{s} + k_\infty s + \sum_{j=1}^m \frac{2k_{mj} s}{s^2 + \omega_m^2}$$

$$= \frac{N(s)}{D(s)}, \quad y(-s) = -y(s)$$

$$= -\frac{N(-s)}{D(-s)}$$

$$y_1(s) = \frac{2k_1 s}{s^2 + \omega_1^2}$$

$$z_1(s) = \frac{s^2 + \omega_1^2}{2k_1 s} = \frac{s}{2k_1} + \frac{1}{\frac{2k_1 s}{\omega_1^2}}$$

$$\frac{1}{h_1 C_1} = \frac{2k_1}{2k_1 \omega_1^2} = \omega_1^2 \Rightarrow \omega_1 = \sqrt{h_1 C_1}$$

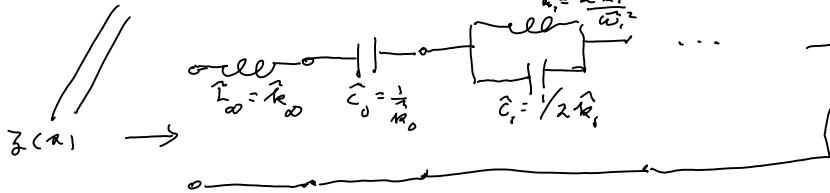
Ex:  $y(s) = \frac{3}{s} + 2s + \frac{6s}{s^2 + 4} = \frac{3(s^2 + 4) + 2s(s^2 + 4) + (6s)}{s(s^2 + 4)} = \frac{2s^4 + 13s^2 + 12}{s(s^2 + 4)}$

$= \frac{\text{Even in } s}{\text{odd in } s} = \frac{\text{Even in } s}{s \times \text{Even in } s}$

as \$z(s)\$ is LBR if \$y(s)\$ is, can use same idea

$$z(s) = \hat{k}_\infty s + \frac{\hat{k}_0}{s} + \sum_{j=1}^m \frac{2\hat{k}_j s}{s^2 + \hat{\omega}_j^2}$$

$$z_j = \frac{2\hat{k}_j s}{s^2 + \hat{\omega}_j^2} \Rightarrow y_j = \frac{s^2 + \hat{\omega}_j^2}{2\hat{k}_j s} = \frac{s}{2\hat{k}_j} + \frac{1}{\frac{2\hat{k}_j s}{\hat{\omega}_j^2}}$$



1st Foster

Ex:  $y(s) = \frac{2s^4 + 13s^2 + 12}{s(s^2 + 4)} = \frac{3}{s} + 2s + \frac{2k_1 s}{s^2 + 4}$

$$s^4 + \frac{13}{2}s^2 + 6 \Rightarrow s_{1,2} = -\frac{13}{4} \pm \frac{1}{2} \sqrt{\left(\frac{13}{2}\right)^2 - 4 \times 6}$$

$$= -\frac{13}{4} \pm \frac{1}{4} \sqrt{13^2 - 16 \times 6}$$

$$= -\frac{13}{4} \pm \frac{1}{4} \sqrt{169 - 96}$$

$$= -\frac{13}{4} \pm \frac{1}{4} \sqrt{73}$$

13	16
13	36
39	6
13	
169	
-96	
73	

$$z(s) = \frac{s(s^2 + 4)}{2\left(s^2 + \frac{13}{4} + \frac{1}{4}\sqrt{73}\right)\left(s^2 + \frac{13}{4} - \frac{1}{4}\sqrt{73}\right)} = \frac{k_1 s}{s^2 + \frac{13}{4} + \frac{1}{4}\sqrt{73}} + \frac{k_2 s}{s^2 + \frac{13}{4} - \frac{1}{4}\sqrt{73}}$$

1st Foster



2nd Foster



these differ but have the same input \$y\$ & \$z = 1/4\$

use the minimum numbers of \$L\$ & \$C\$ = 4 total = degree of \$y\$ = degree of \$z(s)\$ = highest power of \$s\$ in \$y\$ & \$z\$.

