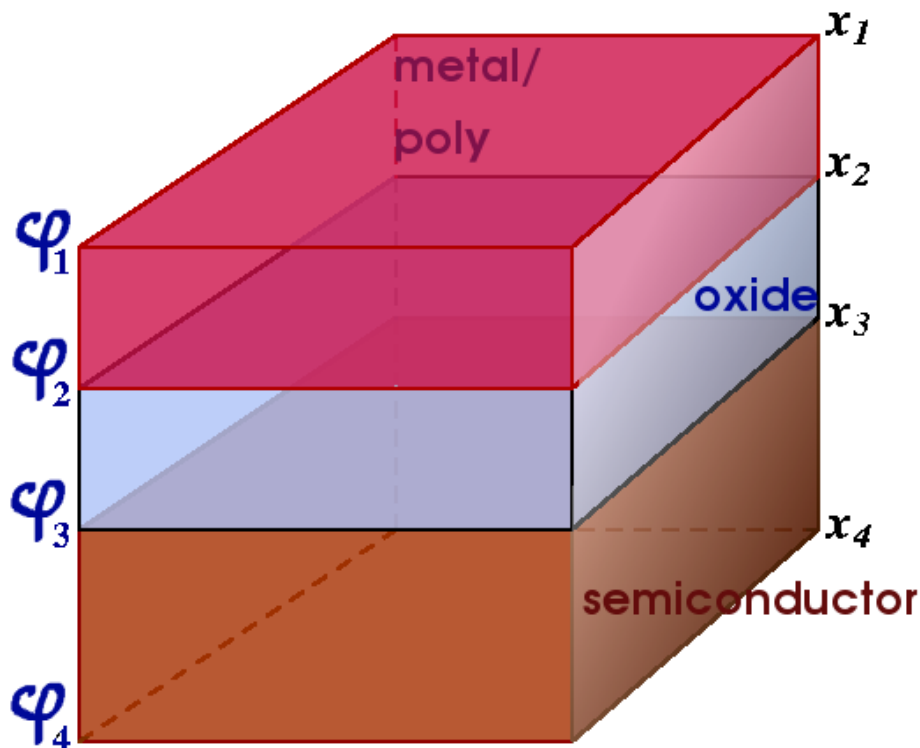


## Derivation of MOSFET Threshold Voltage from the MOS Capacitor

ESEE 313 Notes  
Prof. Neil Goldman

Threshold voltage is the voltage applied between gate and source of a MOSFET that is needed to turn the device on for linear and saturation regions of operation. The following analysis is for determining the threshold voltage of an N-channel MOSFET (also called an N-MOSFET). The analysis is performed with a MOS capacitor like the one shown below.



### Structure

- MOS Capacitor: Top layer is N-type polysilicon or metal. For this derivation we will assume it is heavily doped N-type polysilicon with doping  $N_D$
- $SiO_2$  (oxide) insulation sandwiched between two conductors
- The bottom layer is P-type semiconductor of doping  $N_A$

We will relate the built-in potential of the device to the voltage drop across the three layers. Since we have not added any external voltages, the total drop will be due to the built-in potential only. The total potential drop is the sum of the drops over different layers.

$$\varphi_1 - \varphi_4 = (\varphi_1 - \varphi_2) + (\varphi_2 - \varphi_3) + (\varphi_3 - \varphi_4) \quad (1)$$

$$(\varphi_1 - \varphi_2) = \varphi_{\text{poly}} = \text{drop across poly} \quad (\varphi_{\text{poly}} = 0) \quad (2)$$

$$(\varphi_2 - \varphi_3) = \varphi_{ox} = \text{drop across oxide (3)}$$

$$(\varphi_3 - \varphi_4) = \varphi_{Si} = \text{drop across P-type silicon (4)}$$

Define  $\varphi_1 = \varphi_n$ ,  $\varphi_4 = \varphi_p$ ,  $\varphi_3 = \varphi_s$  &  $\varphi_{bi} = \varphi_n - \varphi_p$  (5), where s=surface between p-Si and SiO<sub>2</sub>

$$\varphi_p = -V_{Th} \ln \frac{N_A}{n_i} = \text{built in potential on P-substrate (6)}$$

$\varphi_s$  is the potential at the surface between the p-Si and SiO<sub>2</sub>

Summing all the drops we write the built in potential as

$$\varphi_{bi} = \varphi_n - \varphi_p = \varphi_{ox} + \varphi_{Si} = \varphi_{ox} + (\varphi_s - \varphi_p) = \text{total potential drop. (7)}$$

Now we know the built in voltage in terms of two unknowns  $\varphi_{ox}$  &  $\varphi_{Si}$ . Our goal now will be to reduce the number of unknowns, and eventually determine the built in voltage as a function of the surface potential  $\varphi_s$  only.

We now need to find  $\varphi_{ox}$ , we know

$$\varphi_{ox} = \int_{x_2}^{x_3} E_{ox} dx = E_{ox} (x_3 - x_2) = E_{ox} t_{ox} \quad (8)$$

( $E_{ox}$  = constant because there is no charge in oxide)

Now need to find  $E_{ox}$ . So the next ten or so lines of equations are written to determine  $E_{ox}$ .

Start by using Gauss' law

$$\epsilon_{ox} E_{ox} \Big|_{x=x_3} = \epsilon_{Si} E_{Si} \Big|_{x=x_3} \quad (9)$$

We can calculate  $E_{Si}$  at interface

Use depletion approximation to calculate  $E_{Si}$ . In the silicon under the oxide the p-type material is depleted due to the built in potential. Thus, there is charge there due to acceptor ions since the holes are absent. Use the depletion approximation to solve for the field in the silicon due to the ionized acceptors. The width of the depletion region is  $w_D$ .

Starting with the Poisson equation, while using the approximation that  $n=0$ ,  $p=0$  in the depletion region, we have:

$$\frac{dE_{Si}}{dx} = \frac{-q}{\epsilon_{Si}} N_A \quad (10)$$

$$E_{Si}(x) = \frac{q}{\epsilon_{Si}} N_A (w_D + x_3 - x) \quad (11)$$

$$E_{Si}(x_3) = \frac{q}{\epsilon_{Si}} N_A w_D \quad (12)$$

$$\varphi_{Si} = \int_{x_3}^{x_4} E_{Si}(x) dx \quad (13)$$

$$= \frac{1}{2} (\text{base})(\text{height}) \text{ of depletion region} = \frac{1}{2} E_{Si}(x_3) w_D = \frac{1}{2} \left( \frac{q}{\epsilon_{Si}} N_A w_D \right) w_D \quad (14)$$

Solve for  $w_D$

$$w_D = \left[ \frac{2\epsilon_{Si}}{qN_A} \varphi_{Si} \right]^{1/2} = \left[ \frac{2\epsilon_{Si}}{qN_A} (\varphi_s - \varphi_p) \right]^{1/2} \quad (15)$$

Finally, substituting  $w_D$  of (15) into equation (12), we have some usable expression for field in terms of the surface potential:

$$\begin{aligned} E_{ox} &= \frac{\epsilon_{Si}}{\epsilon_{ox}} E_{Si} = \frac{\epsilon_{Si}}{\epsilon_{ox}} \frac{q}{\epsilon_{Si}} N_A \left[ \frac{2\epsilon_{Si}}{qN_A} (\varphi_s - \varphi_p) \right]^{1/2} \\ &= \frac{1}{\epsilon_{ox}} \left[ 2\epsilon_{Si} q N_A (\varphi_s - \varphi_p) \right]^{1/2} \end{aligned} \quad (16)$$

Note that all terms in (16) except for the surface potential are determined by the manufacturer, or are material constants.

Finally we can now find  $\varphi_{ox}$

$$\varphi_{ox} = E_{ox} t_{ox} = \frac{t_{ox}}{\epsilon_{ox}} \left[ 2\epsilon_{Si} q N_A (\varphi_s - \varphi_p) \right]^{1/2} = \frac{1}{C_{ox}} \left[ 2\epsilon_{Si} q N_A (\varphi_s - \varphi_p) \right]^{1/2} \quad (17)$$

where  $C_{ox} = \frac{\epsilon_{ox}}{t_{ox}}$  is the capacitance of the oxide capacitance of the MOSCAP per unit area.

An important point to notice here is that  $\varphi_{ox}$  is now written in terms of  $\varphi_s$ , so we have reduced the number of unknowns to one.

Now back to original equation

$$\varphi_{bi} = \varphi_n - \varphi_p = \varphi_{ox} + \varphi_{Si} = \frac{1}{C_{ox}} \left[ 2\epsilon_{Si} q N_A (\varphi_s - \varphi_p) \right]^{1/2} + (\varphi_s - \varphi_p) \quad (18)$$

$\varphi_{bi}$  is also called  $\varphi_{ms} = -V_{FB}$

Note that all terms in this equation are now known except for  $\varphi_s$

### Include Gate Voltage

Now, instead of working with just the built in potential, we add a voltage  $V_G$  to the gate of the MOS capacitor. Now the equation for the total electrostatic potential drop across the MOS capacitor is:

$$V_G + \varphi_{bi} = \varphi_{ox} + \varphi_{Si} = \varphi_{ox} + (\varphi_s - \varphi_p) = \text{total potential drop. (19)}$$

Of course, since we have added  $V_G$ , values for  $\varphi_{ox}$  and  $\varphi_{Si}$  will change. However, if we look back at equation (18), we see that the only parameter that is not already determined in that equation is the surface potential  $\varphi_s$ .  $V_G$  will cause the potential at the interface  $\varphi_s$  to change. So we can conclude that the effect of adding a gate voltage to the MOS capacitor will cause the surface potential to change as shown in equation (20):

$$V_G = \frac{1}{C_{ox}} \left[ 2\varepsilon_{Si} q N_A (\varphi_s - \varphi_p) \right]^{1/2} + \varphi_s - \varphi_p - \varphi_{bi} \quad (20).$$

As shown in equation (20), changing the gate voltage will change the surface potential. This will ultimately change the concentration of electrons at the Si-SiO<sub>2</sub> interface, and eventually, for a MOSFET adjust the drain current for the devices.

### THRESHOLD VOLTAGE

**Threshold Voltage Definition:**  $V_{TH}$  is the value of  $V_G$  that will cause the interface potential to be equal in magnitude and opposite in sign to the substrate potential  $\varphi_p$ . Physically this means that there would now be a mobile electron concentration at the surface that is equal in magnitude to the mobile hole concentration in the p-substrate. When this happens we say that the surface is **INVERTED**, and the electron channel at the surface is called the **inversion layer**.

**So define threshold voltage as**

$$V_G = V_{TH} \quad \text{when} \quad \varphi_s = -\varphi_p \quad (21)$$

Substituting in  $V_{TH}$  and  $\varphi_s$ , we obtain the following expression for **the threshold** voltage in n-channel MOSFETs.

$$V_{TH} = \frac{1}{C_{ox}} \left[ 4\varepsilon_{Si} q N_A |\varphi_p| \right]^{1/2} + 2|\varphi_p| - \varphi_{bi}$$